

Introduction

- Multicast beamforming Uses multiple transmit antennas to send the same data (e.g., video) stream to a group of users @ given QoS.
- Channel State Information at the Transmitter (CSIT) required; can be instantaneous, long-term, or statistical.
- Maximum multicast rate determined by weakest link (smallest SNR).
- Multicast beamformer design formulated as an optimization problem
 - i) Max-Min (SNR) under power constraint
 - ii) Min power under min SNR constraints
- **Non-convex and NP-hard**

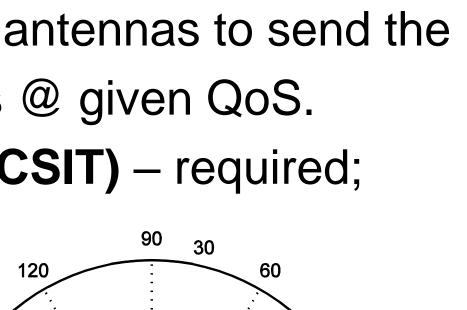
Related Work

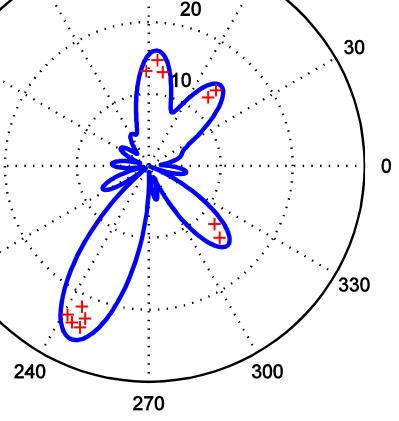
- **SDR-G** [Sidiropoulos *et al.*, 2006]. Semi-Definite Relaxation (SDR) followed by Gaussian randomization.
- Successive Linear Approximation (SLA) [Tran et al., 2014] Iteratively solve sequence of QP problems.
- Alternating Maximization (AM) [Demir et al., 2014] Iteratively solve sequence of SDP problems; high complexity.
- Lozano's algorithm [Lozano, 2007] Step in the direction of the SNR gradient of the bottleneck user, then scale. May not converge; limit cycles [Matskani *et al.*, 2009].
- Damped Lozano's algorithm with Lopez initialization (dLLI) [Matskani *et al.*, 2009].
- **QR algorithm** [Abdelkader *et al.*, 2010] Uses channel orthogonalization based on QR decomposition.
- Multiplicative (MU) Algorithm [Gopalakrishnan et al., 2015] Uses proportional-fairness as a surrogate for max-min fairness, takes into account SNR gradient of all users, achieves best performancecomplexity tradeoff.

System Model

- Single-group multicast cell, single Transmitter (Tx) with N antennas serving M single receive antenna users. Perfect CSIT assumed.
- Received signal at user m: $y_m = \mathbf{h}_m^H \mathbf{w} x + z_m, \forall m \in \mathcal{M} = \{1, \cdots, M\}$
- SNR at user m: $\frac{|\mathbf{h}_m^H \mathbf{w}|^2}{\sigma^2} = \mathbf{w}^H \tilde{\mathbf{R}}_m \mathbf{w}, \forall m \in \mathcal{M}$

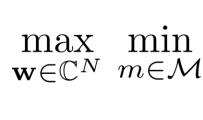
A FAST APPROXIMATION ALGORITHM FOR SINGLE-GROUP MULTICAST BEAMFORMING WITH LARGE ANTENNA ARRAYS **Aritra Konar and Nicholas D. Sidiropoulos** University of Minnesota, USA





Problem Statement

Max-min fair beamformer design problem



- Constraints can be either sum power or per-antenna

Problem Reformulation

min-max problem.

 $\max_{\mathbf{w}\in\mathbb{C}^N}\min_{m\in\mathcal{M}}\mathbf{w}^H\tilde{\mathbf{R}}_m\mathbf{w} \text{ s.t. } \mathbf{w}\in\mathcal{F} \Leftrightarrow \max_{\bar{\mathbf{w}}\in\mathbb{R}^{2N}}\min_{m\in\mathcal{M}}\bar{\mathbf{w}}^T\bar{\mathbf{R}}_m\bar{\mathbf{w}} \text{ s.t. } \bar{\mathbf{w}}\in\bar{\mathcal{F}}$ $\Leftrightarrow \min_{\bar{\mathbf{w}} \in \mathbb{R}^{2N}} \max_{m \in \mathcal{M}} \bar{\mathbf{w}}^T \bar{\mathbf{S}}_m \bar{\mathbf{w}} \quad \text{s.t.} \ \bar{\mathbf{w}} \in \bar{\mathcal{F}}$

where $\bar{\mathbf{w}} := [\mathbf{w}_r^T, \mathbf{w}_i^T]^T \in \mathbb{R}^{2N}, \, \mathbf{w}_r = \operatorname{Re}\{\mathbf{w}\}, \, \mathbf{w}_i = \operatorname{Im}\{\mathbf{w}\}$

SLA for max-min fairness

- Define $u_m(\bar{\mathbf{w}}) := \bar{\mathbf{w}}^T \bar{\mathbf{S}}_m \bar{\mathbf{w}}, \ \forall \ m \in \mathcal{M}$
- **Linearization** about $\bar{\mathbf{w}} = \bar{\mathbf{w}}^{(k)}$ yields
- Furthermore, $\max_{m \in \mathcal{M}} \mathbf{a}_m^{(k)T} \bar{\mathbf{w}} + b_m^{(k)} \ge \max_{m \in \mathcal{M}} u_m(\bar{\mathbf{w}})$
- Algorithm:

 $\bar{\mathbf{w}}^{(k+1)} = \arg\min_{\bar{\mathbf{w}}\in\bar{\mathcal{F}}} \max_{m\in\mathcal{M}} \mathbf{a}_m^{(k)T} \bar{\mathbf{w}} + b_m^{(k)}$

Generates sequence of iterates with monotonically non-increasing cost.

Solving SLA sub-problem

- complexity, fast convergence.
- Subgradient methods: low per-iteration complexity, slow convergence.
- methods.

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$$\mathbf{w}^H \tilde{\mathbf{R}}_m \mathbf{w} \text{ s.t. } \mathbf{w} \in \mathcal{F}$$

Goal: Design first-order based SLA algorithm for max-min fairness

Convert problem to real domain and then equivalently reformulate as

 $\bar{\mathbf{R}}_m := \begin{bmatrix} \operatorname{Re}\{\tilde{\mathbf{R}}_m\} & -\operatorname{Im}\{\tilde{\mathbf{R}}_m\} \\ \operatorname{Im}\{\tilde{\mathbf{R}}_m\} & \operatorname{Re}\{\tilde{\mathbf{R}}_m\} \end{bmatrix} \in \mathbb{R}^{2N \times 2N}, \ \bar{\mathbf{S}}_m := -\bar{\mathbf{R}}_m, \ \forall \ m \in \mathcal{M}$

 $u_m(\bar{\mathbf{w}}) \le u_m(\bar{\mathbf{w}}^{(k)}) + \nabla u_m(\bar{\mathbf{w}}^{(k)})^T(\bar{\mathbf{w}} - \bar{\mathbf{w}}^{(k)}) = \mathbf{a}_m^{(k)T} \bar{\mathbf{w}} + b_m^{(k)}, \forall \ m \in \mathcal{M}$

Interior point methods: standard approach, high per-iteration

Linearized Alternating Direction Methods of Multipliers (L-ADMM):

variant of classical ADMM algorithm, low per-iteration complexity, convergence order of magnitude faster compared to subgradient

